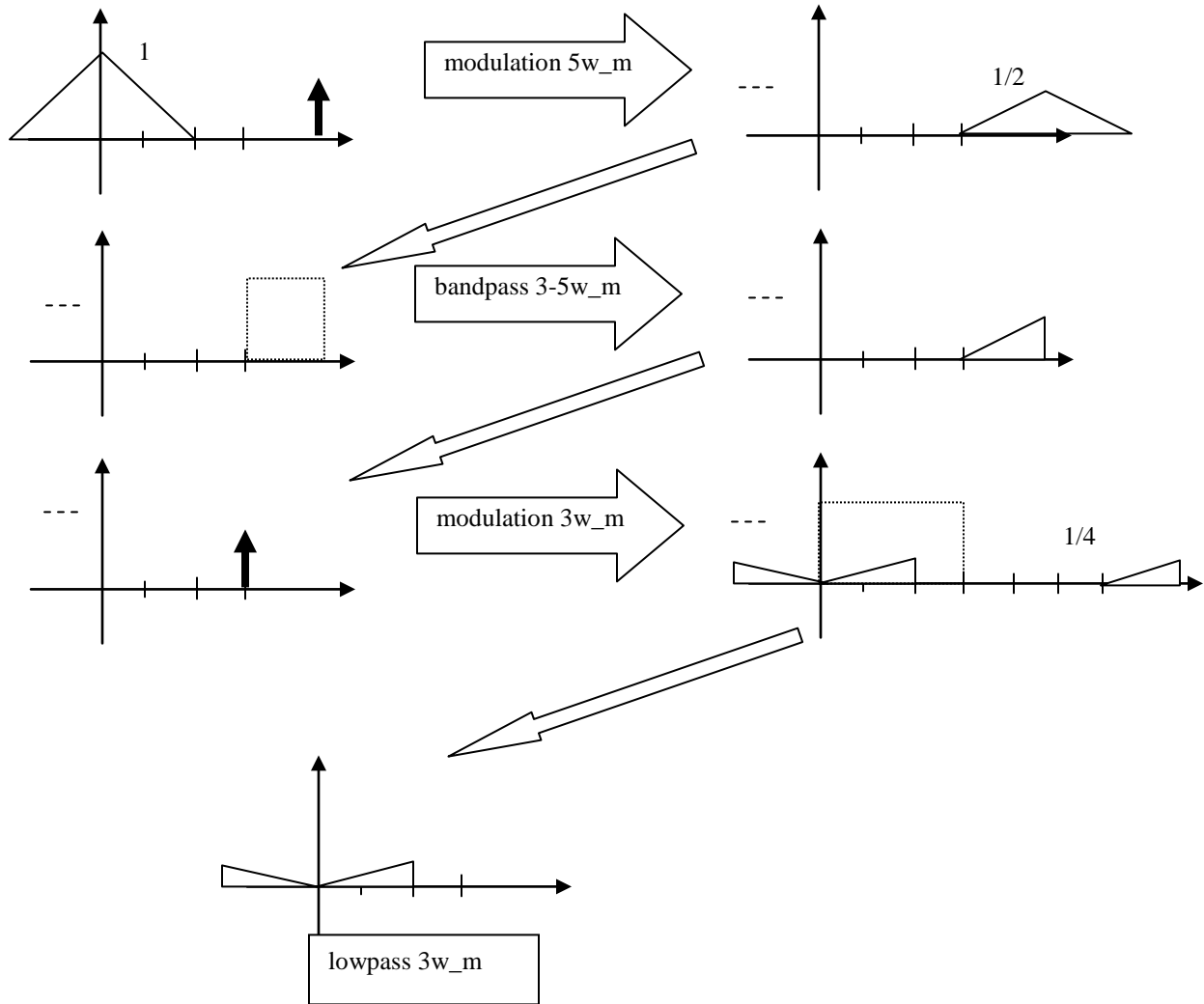


**Problem 1:**

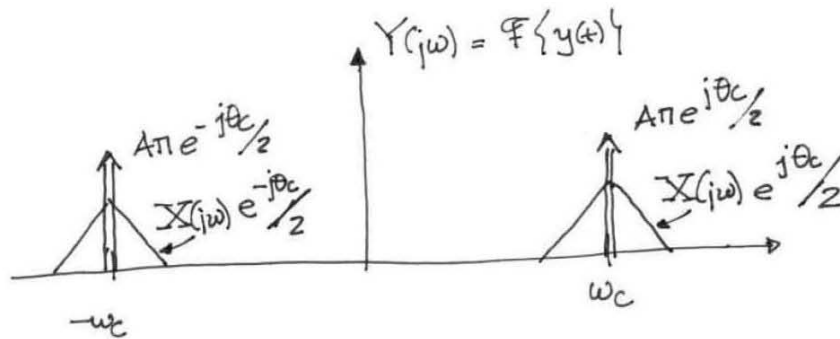
Do Problem 8.22 from the textbook.



**Problem 2:**

Do Problem 8.26 from the textbook.

Pr 8.26



$$\mathcal{F}\{y(t) \cos \omega_c t\} = \begin{aligned} & \text{Diagram 1: } \omega\text{-axis with pulses at } -\omega_c \text{ and } \omega_c. \text{ Labels: } 1/4 X(j\omega) e^{j\theta_c} + 1/2 A\pi e^{j\theta_c} \delta(\omega) \\ & \text{Diagram 2: } \omega\text{-axis with pulses at } -\omega_c \text{ and } \omega_c. \text{ Labels: } 1/4 X(j\omega) e^{-j\theta_c} + 1/2 A\pi e^{-j\theta_c} \delta(\omega) \end{aligned}$$

$$= \begin{aligned} & \text{Diagram 3: } \omega\text{-axis with a central pulse at } 0. \text{ Labels: } A\pi \cos \theta_c, \frac{1}{2} X(j\omega) \cos \theta_c \end{aligned}$$

After lowpass  $\rightarrow$

$$2A\pi \cos \theta_c \quad X(j\omega) \cos \theta_c = \mathcal{F}\{(x(t)+A) \cos \theta_c t\}$$

Similarly for the  $\sin \omega_c t$  modulation, (the side lobes are different)

$\rightarrow \dots \rightarrow$  after lowpass

$$2A\pi \sin \theta_c \quad X(j\omega) \sin \theta_c = \mathcal{F}\{(x(t)+A) \sin \theta_c t\}$$

Hence, the outputs of the low-pass filters are  $(x(t)+A) \cos \theta_c$  and  $(x(t)+A) \sin \theta_c$ . Squaring and adding we obtain  $[x(t)+A]^2 [\cos^2 \theta_c + \sin^2 \theta_c] = [x(t)+A]^2$ . Therefore,  $r(t) = x(t) + A$ . (Note:  $x(t)+A > 0$ )

**Problem 3: (8.49 of textbook)**

(a)  $s(t)$  periodic so its Fourier transform is computed through the Fourier series expansion. That is,

$$s(t) = \sum a_k e^{jk \frac{2\pi}{T} t}, \quad \omega_o = \frac{2\pi}{T}$$

$$S(j\omega) = 2\pi \sum a_k \delta\left(\omega - k \frac{2\pi}{T}\right), \quad a_k = \frac{\sin k\omega_o T_1}{k\pi}, \quad T_1 = \frac{T}{4} \Rightarrow a_k = \frac{\sin k\pi/2}{k\pi}$$

$$a_0 = 1/2$$

(The  $a_k$  is zero for even  $k$ , other than 0.)

The modulated (chopped)  $x(t)$  has Fourier transform

$$X_s(j\omega) = \frac{1}{2\pi} X(j\omega) * S(j\omega) = \sum a_k X\left(j\omega - jk \frac{2\pi}{T}\right)$$

yielding the band-pass filtered-chopped signal, say  $v(t)$

$$\begin{aligned} V(j\omega) &= H_1(j\omega) \sum a_k X\left(j\omega - jk \frac{2\pi}{T}\right) = Aa_1 X\left(j\omega - j \frac{2\pi}{T}\right) + Aa_{-1} X\left(j\omega + j \frac{2\pi}{T}\right) \\ &= \frac{A}{\pi} X\left(j\omega - j \frac{2\pi}{T}\right) + \frac{A}{\pi} X\left(j\omega + j \frac{2\pi}{T}\right) \end{aligned}$$

The maximum allowable frequency content in  $x(t)$  for this expression to be valid is  $\omega_M < \pi/T$ .

Next,  $V(j\omega)$  is re-modulated (chopped) and low-pass filtered. The modulation by  $s(t)$  produces two replicas of  $X(j\omega)$  at 0, each multiplied by  $1/\pi$ , yielding a total coefficient of  $2A/\pi^2$ . The rest of the replicas are at  $k2\pi/T$  that are filtered out by the  $H_2$  low-pass filter. So,

$$Y(j\omega) = H_2(j\omega) V_s(j\omega) = \frac{2A}{\pi^2} X(j\omega)$$

with the same condition on the maximum frequency in  $x(t)$ ,  $\omega_M < \pi/T$ .

(b) From the last expression, the equivalent gain of the overall system is  $2A/\pi^2$ .